±65V Fault Protected 500kpbs/20Mbps Half-Duplex RS-485/RS-422 Transceivers

General Description

The MAX14775E/MAX14776E fault-protected RS-485/ RS-422 transceivers feature ±65V protection for overvoltage signal faults on communication bus lines, ensuring communication in harsh industrial environments. Each device contains one driver and one receiver and operates over the 3V to 5.5V supply range. The MAX14775E is optimized for high-speed data rates up to 20Mbps. The MAX14776E features slew-rate limited outputs for data rates up to 500kbps.

These transceivers are optimized for robust communication in noisy environments. A large 200mV (typ) hysteresis on receiver inputs ensure for high noise rejection and a failsafe feature guarantees a logic-high on the receiver output when the inputs are open or shorted. Driver outputs are protected against short-circuit conditions.

The MAX14775E/MAX14776E receivers feature a 1/3unit load input impedance, allowing up to 100 transceivers on a bus.

The MAX14775E/MAX14776E are available in 8-pin SOIC and 8-pin TDFN-EP packages and operate over the -40°C to +125°C temperature range.

Benefits and Features

- Integrated Protection Ensures for Robust
 Communication
 - ±65V Fault Protection Range on Driver Outputs/ Receiver Inputs
 - ±25V Common Mode Range on the Receiver Inputs
 - Large Receiver Hysteresis Increases Noise Tolerance
 - Hot-Swap Protection
 - Thermal Shutdown
- High-Performance Transceiver Enables Flexible Designs
 - Compliant with RS-485 EIA/TIA-485 Standard
 - 20Mbps (MAX14775E)/500kbps (MAX14776E) Maximum Data Rate
 - 3V to 5.5V Supply Range
 - Up to 100 Devices on the Bus

Applications

- Industrial Field Bus Networks
- Motion Controllers
- HVAC

Ordering Information appears at end of data sheet.

Selector Guide

PART NUMBER	MAX DATA RATE	PIN-PACKAGE
MAX14775EASA+	20Mbps	8 SOIC
MAX14775EATA+	20Mbps	8 TDFN-EP
MAX14776EASA+	500kbps	8 SOIC
MAX14776EATA+	500kbps	8TDFN-EP



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Absolute Maximum Ratings

(All voltages referenced to GND)

V _{CC}	0.3V to +6V
RÕ	0.3V to (V _{CC} + 0.3V)
DE, DI, RE	-0.3V to +6V
A, B (I _{MAX} = ±1mA)	70V to +70V
Short-Circuit Duration (RO, A, B)	Continuous

Continuous Power Dissipation (T _A = +70°C)
8-pin SOIC (derate 7.60mW/°C above +70°C)606.1mW
8-pin TDFN (derate 24.4mW/°C above +70°C)1951.2mW
Operating Temperature Range40°C to +125°C
Junction Temperature+150°C
Storage Temperature Range65°C to +150°C
Lead Temperature (soldering, 10s)+300°C
Soldering Temperature (reflow)+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Package Thermal Characteristics (Note 1)

SOIC

Junction-to-Ambient Thermal Resistance (θ_{JA})......132°C/W Junction-to-Case Thermal Resistance (θ_{JC}).......38°C/W

TDFN

- Junction-to-Ambient Thermal Resistance (θ_{JA}).......41°C/W Junction-to-Case Thermal Resistance (θ_{JC})..........8°C/W
- Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

DC Electrical Characteristics

(V_{CC} = 3.0V to 5.5V, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at V_{CC} = 3.3V and T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER						
Supply Voltage	V _{CC}		3.0		5.5	V
Supply Current	ICC	DE = high, \overline{RE} = low, no load, no 3 5. switching		5.3	mA	
Shutdown Supply Current	I _{SH}	DE = high, \overline{RE} = low		4		μA
Shutdown Short-Circuit Supply Current	ISHDN_SHRT	A or B shorted to \pm 65V, DE = high, RE = low		240		μA
DRIVER		·				
Differential Driver Output	IVodi	$R_L = 54\Omega$, Figure 1a	1.5			v
	IVODI	R _L = 100Ω, Figure 1a	2.0			
Change in Magnitude of Differ- ential Driver Output Voltage	ΔV _{OD}	R _L = 100Ω or 54Ω, Figure 1a (Note 3)	-0.2		+0.2	V
Driver Common-Mode Output Voltage	Voc	R _L = 100Ω or 54Ω, Figure 1a		V _{CC} / 2	3	v
Change in Magnitude of Common-Mode Voltage	ΔV _{OC}	R _L = 100Ω or 54Ω, Figure 1a (Note 3)	-0.2		+0.2	v
Single-Ended Driver Output Voltage High	VOH	A and B outputs, output is high, ISOURCE = 3mA	V _{CC} -0.2			v
Single-Ended Driver Output Voltage Low	V _{OL}	A and B outputs, output is low, ISINK = 3mA			0.2	v
Driver Short-Circuit Output Current	IOSD1	$\begin{array}{l} -65 V \leq V_{A \text{ or }} V_{B} < 0 V \text{ or } V_{CC} < V_{A \text{ or}} \\ V_{B} \leq +65 V \text{ (Note 4)} \end{array}$			200	mA

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DC Electrical Characteristics (continued)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
Average Driver Short-Circuit Output Current	IOSD2	$0V \le V_{A \text{ or }} V_B \le V_{CC}$				250	mA
RECEIVER							
Input Current (A. D)	1. 1-	DE = low,	V _{CM} = +12V			+280	
Input Current (A, B)	I _A , I _B	$0V \le V_{CC} \le 5.5V$ $V_{CM} = -7V$ -2		-200			μA
Receiver Input Resistance	R _{IN}	$-7V \le V_{CM} \le +12V$	/	38		-	kΩ
Common Mode Voltage Range	V _{CM}			-25		+25	V
Receiver Differential Threshold Voltage Rising	V _{THH}	$-25V \le V_{CM} \le +25$	V	+40		+200	mV
Receiver Differential Threshold Voltage Falling	VTHL	$-25V \le V_{CM} \le +25$	V	-200		-40	mV
Receiver Input Hysteresis	ΔV_{TH}	V_{CM} = 0V, time from $< t_{D}$ FS	om last transition		250		mV
Differential Input Fail-safe Threshold	VTH_FSH	$25V \le V_{CM} \le +25V$ transition > tD_FS		-40		+40	mV
Differential Input Capacitance	C _{A,B}	Measured between	n A and B, f = 1MHz		50		pF
LOGIC OUTPUTS (RO)							1
RO Output Logic High Voltage	VOH	ISOURCE = 3mA,	(V _A - V _B) ≥ +200mV	V _{CC} -0.4			V
RO Output Logic Low Voltage	V _{OL}	I _{SINK} = 3mA, (V _A	- V _B) < +200mV			0.4	V
RO Leakage Current	lozr	\overline{RE} = high, 0V \leq V	RO ≤ VCC	-1		+1	μA
RO Short-Circuit Current	IOSR	$0V \le V_{RO} \le V_{CC}$			70		mA
LOGIC INPUTS (DE, DI, RE)							
Input Logic High Voltage	VIH			0.67 x V _{CC}			v
Input Logic Low Voltage	VIL					0.33 x V _{CC}	V
Input Hysteresis	V _{HYS}				100		mV
Input Leakage Current	IIN			-1		+1	μA
Input Impedance on First Transition	RIN_FT	DE, RE		1		10	kΩ
PROTECTION		-1					
Thermal-Shutdown Threshold	T _{SHDN}	Temperature rising			+162		°C
Thermal-Shutdown Hysteresis	T _{HYST}				12		°C
ESD Protection		Human Body Mode	el		±8		
(A, B Pins to GND)		IEC 61000-4-2- Contact Discharge			±5		kV
ESD Protection (All Other Pins)		Human Body Mode	el		±2		kV
Fault Protection Range (A, B Pins to GND)				-65		+65	v

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Switching Electrical Characteristics (MAX14775E)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DRIVER						
Driver Propagation Delay	^t DPLH, ^t DPHL	R_L = 54 Ω , C_L = 50pF, Figure 2 and Figure 3			40	ns
Differential Driver Output Skew [^t DPLH - ^t DPHL]	^t DSKEW	R_L = 54 Ω , C_L = 50pF, Figure 2 and Figure 3 (Note 7)			9	ns
Driver Differential Output Rise or Fall Time	tLH, tHL	R_L = 54 Ω , C_L = 50pF, Figure 2 and Figure 3 (Note 7)		8	15	ns
Maximum Data Rate	DRMAX		20			Mbps
Driver Enable to Output High	^t DZH	R_L = 110 Ω , C_L = 50pF, Figure 4			90	ns
Driver Enable to Output Low	^t DZL	R_L = 110 Ω , C_L = 50pF, Figure 5			90	ns
Driver Enable Time	tD	-20V ≤V _{CM} ≤+25V, 4.5V ≤V _{CC} ≤5.5V, Figure 1a			1000	ns
Driver Disable Time From Low	^t DLZ	R_L = 110 Ω , C_L = 50pF, Figure 5			50	ns
Driver Disable Time From High	^t DHZ	R_L = 110 Ω , C_L = 50pF, Figure 4			50	ns
Driver Enable Time from Shutdown to Output High	^t DLZ(SHDN)	$R_L = 110\Omega$, $C_L = 50pF$, Figure 4 (Note 5)			170	μs
Driver Enable Time from Shutdown to Output Low	^t DHZ(SHDN)	R_L = 110Ω, C_L = 50pF, Figure 4 (Note 5)			170	μs
Time to Shutdown	^t SHDN	(Note 5)	50		800	ns
RECEIVER (Note 6)						
Receiver Propagation Delay	^t RPLH, ^t RPHL	C_L = 15pF, Figure 6 and Figure 7			50	ns
Receiver Output Skew	^t RSKEW	C _L = 15pF, Figure 6 and Figure 7 (Note 7)			5	ns
Receiver Enable to Output High	^t RZH	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8			50	ns
Receiver Enable to Output Low	^t RZL	$R_L = 1k\Omega$, $C_L = 15pF$, S1 closed, Figure 8			50	ns
Receiver Disable Time From Low	^t RLZ	R_L = 1kΩ, C _L = 15pF, S1 closed, Figure 8			50	ns
Receiver Disable Time From High	^t RHZ	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8			50	ns
Receiver Enable from Shutdown to Output Low	^t RLZ(SHDN)	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8 (Note 5)			170	μs
Receiver Enable from Shutdown to Output High	^t RHZ(SHDN)	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8 (Note 5)			170	μs
Time to Shutdown	^t SHDN	(Note 5)	50		800	ns
Delay to Fail-Safe Operation	^t D_FS			10		μs

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Switching Electrical Characteristics (MAX14776E)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
DRIVER							
Driver Propagation Delay	^t DPLH, ^t DPHL	$R_L = 54\Omega, C_L = 50$ Figure 3	$R_L = 54\Omega$, $C_L = 50pF$, Figure 2 and Figure 3			1000	ns
Differential Driver Output Skew ^t DPLH ^{- t} DPHL	^t DSKEW	$R_L = 54\Omega, C_L = 50$ Figure 3 (Note 7)	pF, Figure 2 and			140	ns
Driver Differential Output Rise or Fall Time	tLH, tHL	$R_L = 54\Omega, C_L = 50pF, Figure 2 and Figure 3$	$3V \le V_{CC} \le 3.6V$ $4.5V \le V_{CC} \le 5.5V$	105 105		600 600	ns
Maximum Data Rate	DRMAX			500			kbps
Driver Enable to Output High	tDZH	R _L = 110Ω, C _L = 5	0pF, Figure 4			2500	ns
Driver Enable to Output Low	tDZL	R _L = 110Ω, C _L = 5	0pF, Figure 5			2500	ns
Driver Enable Time	tD	-20V ≤ V _{CM} ≤ +25V Figure 1a	′, 4.5V ≤ V _{CC} ≤5.5V,			3500	ns
Driver Disable Time From Low	^t DLZ	$R_{L} = 110\Omega, C_{L} = 50$	0pF, Figure 5			100	ns
Driver Disable Time From High	^t DHZ	R_L = 110Ω, C_L = 50	0pF, Figure 4			100	ns
Driver Enable Time from Shutdown to Output High	^t DLZ(SHDN)	R_L = 110Ω, C_L = 50pF, Figure 4 (Note 5)				170	μs
Driver Enable Time from Shutdown to Output Low	^t DHZ(SHDN)	$R_L = 110\Omega$, $C_L = 50pF$, Figure 4 (Note 5)				170	μs
Time to Shutdown	^t SHDN	(Note 5)		50		800	ns
RECEIVER (Note 6)							
Receiver Propagation Delay	^t RPLH, ^t RPHL	CL = 15pF, Figure	6 and Figure 7			200	ns
Receiver Output Skew	^t RSKEW	C _L = 15pF, Figure ((Note 7)	6 and Figure 7			30	ns
Receiver Enable to Output High	^t RZH	$R_L = 1k\Omega$, $C_L = 15$ Figure 8				50	ns
Receiver Enable to Output Low	^t RZL	$R_L = 1k\Omega, C_L = 15$ Figure 8				50	ns
Receiver Disable Time from Low	^t RLZ	$R_L = 1k\Omega$, $C_L = 15pF$, S1 closed, Figure 8				50	ns
Receiver Disable Time from High	^t RHZ	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8				50	ns
Receiver Enable from Shutdown to Output High	^t RLZ(SHDN)	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8				170	μs
Receiver Enable from Shutdown to Output Low	^t RHZ(SHDN)	$R_L = 1k\Omega$, $C_L = 15pF$, S2 closed, Figure 8				170	μs
Time to Shutdown	^t SHDN	(Note 5)		50		800	ns
Delay to Fail-Safe Operation	^t D FS				10		μs

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Switching Electrical Characteristics (MAX14776E) (continued)

- **Note 2:** All devices are 100% production tested at $T_A = +25^{\circ}C$. Specifications over temperature are guaranteed by design.
- Note 3: ΔV_{OD} and ΔV_{OC} are the changes in V_{OD} and V_{OC} , respectively, when the DI input changes state.
- **Note 4:** The short-circuit current is 200mA (max) for a short period (35µs, typ). If the short circuit persists, the outputs are then set to high impedance for 300ms (typ).
- **Note 5:** Shutdown is enabled when RE is high and DE is low. If the enable inputs are in this state for less than 50ns, the device is guaranteed not to enter shutdown. If the enable inputs are held in this state for at least 800ns, the device is guaranteed to have entered shutdown.
- Note 6: Capacitive load includes test probe and fixture capacitance.
- Note 7: Guaranteed by design. Not production tested.

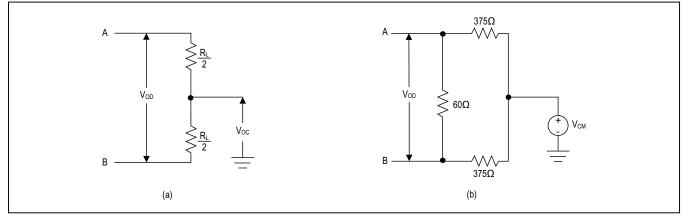


Figure 1. Driver DC Test Load

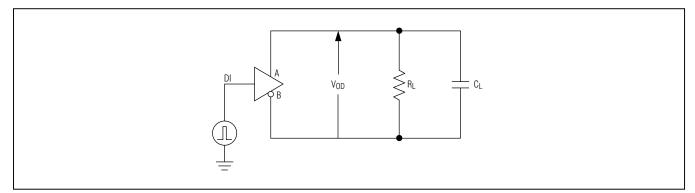


Figure 2. Driver Timing Test Circuit

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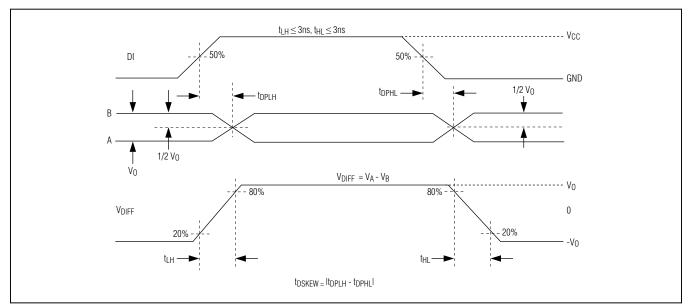


Figure 3. Driver Propagation Delays

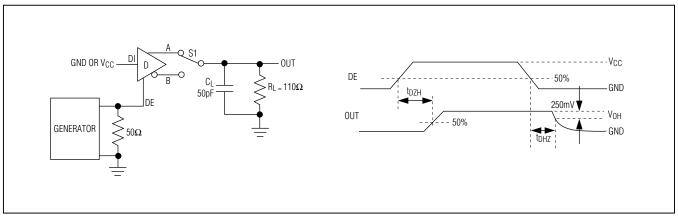


Figure 4. Driver Enable and Disable Times (t_{DHZ}, t_{DZH})

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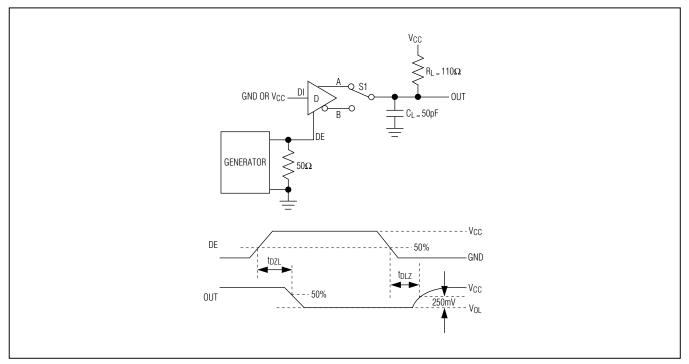


Figure 5. Driver Enable and Disable Times (t_{DZL}, t_{DLZ})

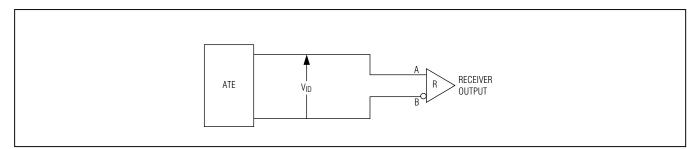


Figure 6. Receiver Propagation Delay Test Circuit

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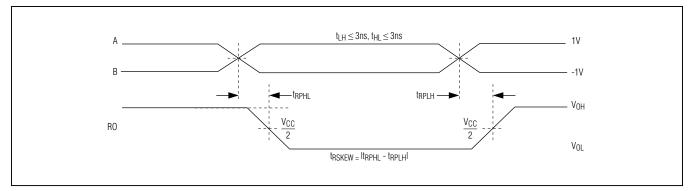


Figure 7. Receiver Propagation Delays

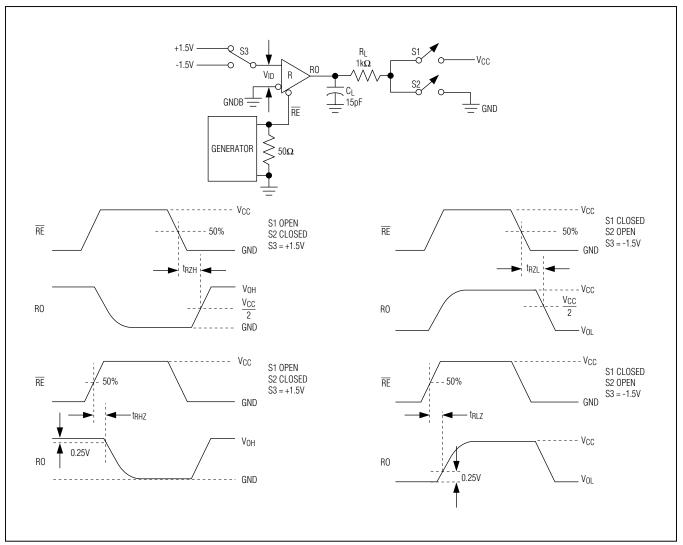
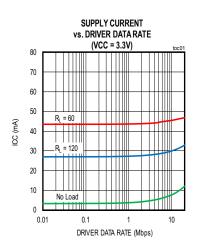


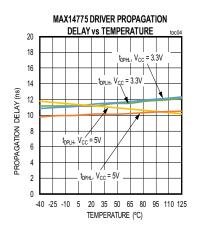
Figure 8. Receiver Enable and Disable Times

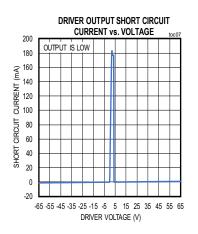
±65V Fault Protected 500kpbs/20Mbps Half-Duplex RS-485/RS-422 Transceivers

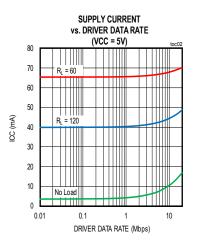
Typical Operating Characteristics

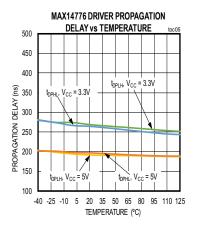
(V_{CC} = 3.3V, T_A = $+25^{\circ}$ C, unless otherwise noted.)

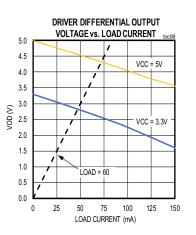


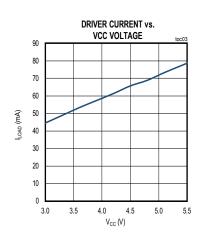


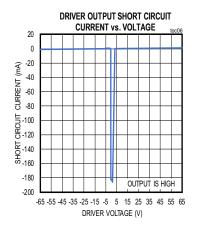


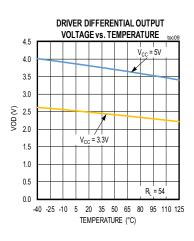








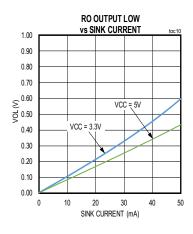


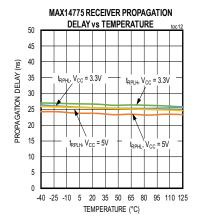


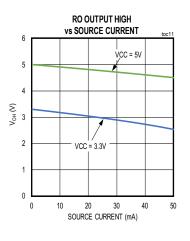
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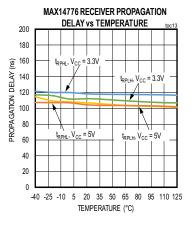
Typical Operating Characteristics (continued)

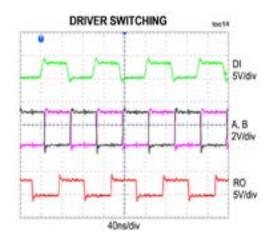
(V_{CC} = 3.3V, T_A = $+25^{\circ}$ C, unless otherwise noted.)





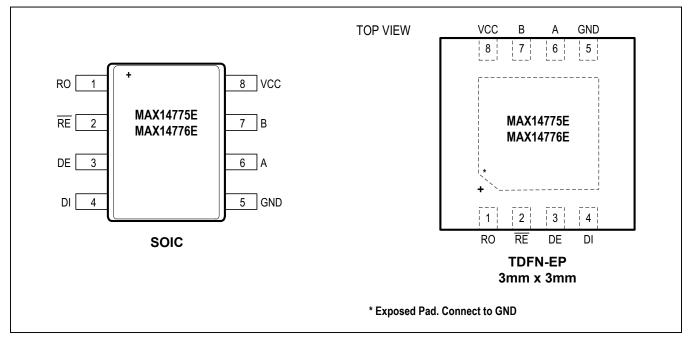






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Pin Configurations



Pin Description

PIN	NAME	FUNCTION
1	RO	Receiver Data Output. See the Function Tables for more information.
2	RE	Receiver Output Enable. Drive \overline{RE} low or connect to GND to enable RO. Drive \overline{RE} high to disable the receiver. RO is high impedance when \overline{RE} is high. Drive \overline{RE} high and DE low to force the IC into low-power shutdown mode.
3	DE	Driver Output Enable. Drive DE high to enable the driver. Drive DE low or connect to GND to disable the driver. Drive DE low and $\overline{\text{RE}}$ high to force the IC into low-power shutdown mode.
4	DI	Driver Input. With DE high, a low on DI forces the noninverting output (A) low and the inverting output (B) high. Similarly, a high on DI forces the noninverting output high and the inverting output low.
5	GND	Ground
6	A	Noninverting Driver Output/Receiver Input
7	В	Inverting Driver Output/Receiver Input
8	Vcc	Power Supply Input. Bypass V_{CC} to GND with a 0.1µF capacitor as close as possible to the device.
_	EP	Exposed Pad. TDFN package only. Connect EP to GND. EP is not intended as the main ground connection.

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Function Tables

	TRANSMITTING						
	INPUTS		OUTI	PUTS			
RE	DE DI		В	A			
Х	1	1	0	1			
X	1	0	1	0			
0	0	Х	High Impedance High Impedance				
1	0	Х	Shutdown. A and B are high impedance.				

Note: X = Don't care.

	RECEIVING							
		INPUTS		OUTPUTS				
RE	DE	(V _A - V _B)	Time from Last A-B Transition	RO				
0	х	≥ +200mV	Always	1				
0	х	-200mV < (V _A - V _B) < +200mV	< tD_FS	Indeterminate. RO is latched to previous value.				
0	х	$-40mV < (V_A - V_B) < +40mV$	> ^t D_FS	1				
0	х	≤ - 200mV	Always	0				
0	х	Open/Shorted	> ^t D_FS	1				
1	1	Х	Х	High impedance				
1	0	Х	Х	Shutdown. RO is high impedance.				

Note: X = Don't care.

±65V Fault Protected 500kpbs/20Mbps Half-Duplex RS-485/RS-422 Transceivers

Detailed Description

The MAX14775E/MAX14776E half-duplex transceivers are optimized for RS-485/RS-422 applications that require up to \pm 65V protection from faults on communication bus lines. These devices contain one differential driver and one differential receiver. The devices feature a 1/3 unit load, allowing up to 100 transceivers on a single bus.

The MAX14775E supports data rates up to 20Mbps. The MAX14776E supports data rates up to 500kbps.

Driver

The driver accepts a single-ended, logic-level input (DI) and transfers it to a differential RS-485 level output on the A and B driver outputs.

Set the driver enable input (DE) low to disable the driver. A and B are high impedance when the driver is disabled.

Receiver

The receiver accepts a differential, RS-485 level input on the A and B inputs and transfers it to a single-ended, logic-level output (RO).

Drive the receiver enable input (\overline{RE}) low to enable the receiver. Driver \overline{RE} high to disable the receiver. RO is high impedance when \overline{RE} is high.

Low-Power Shutdown

Drive DE low and \overline{RE} high for at least 800ns to put the MAX14775E/MAX14776E into low-power shutdown mode. Supply current drops to 20µA when the device is in shutdown mode.

A glitch protection feature ensures that the MAX14775E/ MAX14776E will not accidentally enter shutdown mode due to logic skews between DE and \overline{RE} when switching between transmit and receive modes.

±65V Fault Protection

The driver outputs/receiver inputs of transceivers connected to an industrial RS-485 network often experience faults when shorted to voltages that exceed the -7V to +12V input range specified in the EIA/TIA-485 standard. Under such circumstances, ordinary RS-485 transceivers that have a typical absolute maximum voltage rating of -8V to +12.5V require costly external protection devices which can compromise the RS-485 performance. To reduce system complexity and the need for external protection, the driver outputs/receiver inputs of the MAX14775E/MAX14776E are designed to withstand voltage faults of up to \pm 65V with respect to ground without damage. Protection is guaranteed regardless whether the transceiver is active, in shutdown or without power.

When a fault is detected on A or B, the affected driver output is switched into a high-impedance state. After 300ms (typ), the driver output is re-enabled for $30\mu s$ (typ). If the fault condition persists, the driver output is again disabled. If the fault has been removed, the driver outputs remain on and the transceiver operates normally.

Driving a non-terminated cable may cause the voltage seen at the driver outputs (A or B) to exceed the absolute maximum voltage rating if the DI input is switched during a $\pm 65V$ fault on the A or B pins. Therefore, a termination resistor is recommended in order to maximize the overvoltage fault protection while the DI input is being switched.

If the DI input does not change state while the fault condition is present, the MAX14775E/MAX14776E will withstand up to \pm 65V on the RS-485 inputs, regardless of the termination status of the data cable.

Fail-Safe

The devices' receiver features symmetrical thresholds to improve the duty cycle of the received signal, ensuring that it is 50% when the received signal amplitude is small. Additionally, a high input hysteresis (250mV, typ) increases the resilience to noise on the receiver.

The MAX14775E/MAX14776E also include a fail-safe feature that ensures the receiver output (RO) is high when the receiver inputs are shorted or open, or when they are connected to a differentially terminated transmission line with all drivers disabled for longer than t_D FS (10 μ s, typ).

Hot-Swap Functionality

Hot-Swap Inputs

Inserting circuit boards into a hot, or powered backplane may cause voltage transients on DE, RE, and receiver inputs A and B that can lead to data errors. For example, upon initial circuit board insertion, the processor undergoes a power-up sequence. During this period, the highimpedance state of the output drivers makes them unable to drive the MAX14775E/MAX14776E enable inputs to a defined logic level. Meanwhile, leakage currents of up to 10µA from the high-impedance output, or capacitively coupled noise from V_{CC} or GND, could cause an input to drift to an incorrect logic state. To prevent such a condition from occurring, the MAX14775E/MAX14776E features hot-swap input circuitry on DE and RE to safeguard against unwanted driver activation during hot-swap situations. When VCC rises, an internal pulldown circuit holds DE low and RE high for at least 10µs. After the initial power-up sequence, the internal pulldown/pullup circuitry becomes transparent, resetting the hot-swap tolerable inputs.

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Thermal Shutdown Protection

The MAX14775E/MAX14776E feature thermal-shutdown protection circuitry to protect the device. When the junction temperature exceeds +165°C (typ), the driver outputs are disabled and RO is high impedance. Driver and receiver outputs are re-enabled when the junction temperature falls below 150°C (typ).

Applications Information

100 Transceivers on the Bus

The MAX14775E/MAX14776E transceivers have 0.32unit load receiver, allowing up to 100 MAX14775E/ MAX14776E transceivers connected in parallel on a shared communication line. Connect any combination of these devices, and/or other RS-485 devices, for a maximum of 32 unit loads to the line.

Typical Application

The MAX14775E/MAX14776E half-duplex transceivers are designed for bidirectional data communications on multipoint bus transmission lines. Figure 9 shows a typical network applications circuit. To minimize reflections, the bus should be terminated at the receiver inputs in its characteristics impedance, and stub lengths off the main line should be kept as short as possible.

Power Considerations for the MAX14775E/ MAX14776E

At high data rates, the power dissipation of an RS-485 transceiver can be high. The power dissipation of a halfduplex transceiver is determined by a number of factors, including:

- The data rate
- The time that the driver is transmitting
- The termination impedance
- The power supply voltage

Higher data rates result in higher power dissipation due to switching losses in the transceiver. Switching losses increase even more when capacitance is applied to the A and B pins. External capacitance should be kept to a minimum to help reduce power dissipation at high data rates.

Similarly, the power dissipation in a transceiver is much higher when the driver is transmitting, compared to when the transceiver is receiving. In half-duplex communication, the period of transmission relative to the idle or receiving intervals (i.e., the duty cycle) should be taken into consideration when calculating the average power dissipation.

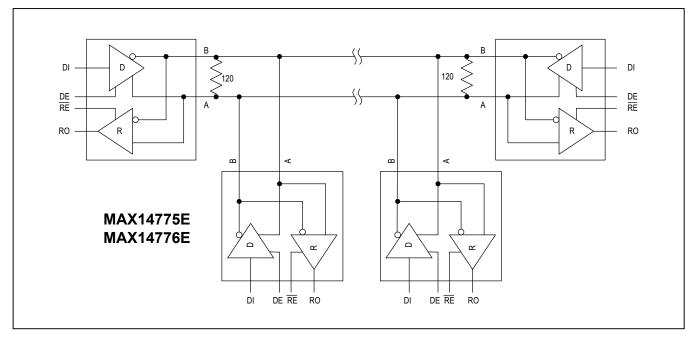


Figure 9. Typical RS-485 Network

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The line termination resistance/impedance determines the driver's load current during transmission and the differential output voltage (V_{OD}) on the driver is determined by the supply voltage. A higher supply voltage results in a larger differential output voltage at the driver driving the line, which in turn results in a higher current draw from the supply (I_{CC}).

The power dissipation in the chip is calculated as the product of supply current times supply voltage, subtracting the power dissipated in the external termination resistor²:

$$P_{DIS} = (V_{CC} \times I_{CC}) - (V_{OD}^2/R_{LOAD})$$

Use the Typical Operation Characteristics to determine the supply current at a given supply voltage and data rate.

For example, assuming a data rate of 20Mbps with a 5V supply on a fully loaded bus ($R_L = 60\Omega$), we can calculate that the power dissipation (at room temperature) is:

 $P_{DIS} = (5V \times 70mA) - (4.3V^2/60\Omega) = 42mW$

Ensure that power dissipation of the transceiver is kept below the value listed in the <u>Absolute Maximum Ratings</u> section to protect the device from entering thermal shutdown or from damage. If the calculated power dissipation nears the specified limits, select a package with a lower thermal resistance which also allows for higher power dissipation.

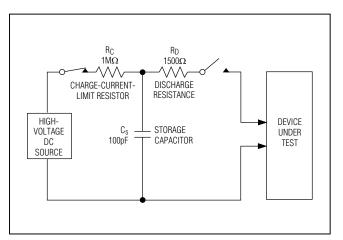


Figure 10. Human Body ESD Test Model

ESD Protection

ESD protection structures are incorporated on all pins to protect against electrostatic discharge encountered during handling and assembly. The driver outputs and receiver inputs of the MAX14775E/MAX14776E have extra protection against static electricity. The ESD structures withstand high ESD in normal operation and when powered down. After an ESD event, the devices keep working without latch-up or damage.

ESD protection can be tested in various ways. The transmitter outputs and receiver inputs of the devices are characterized for protection to the cable-side ground (GNDB) to the following limits:

- ±8kV HBM
- ±5kV using the Contact Discharge method specified in the IEC 61000-4-2

ESD Test Conditions

ESD performance depends on a variety of conditions. Contact Maxim for a reliability report that documents test setup, test methodology, and test results.

Human Body Model (HBM)

<u>Figure 10</u> shows the HBM test model and <u>Figure 11</u> shows the current waveform it generates when discharged in a low-impedance state. This model consists of a 100pF capacitor charged to the ESD voltage of interest, which is then discharged in to the test device through a $1.5k\Omega$ resistor.

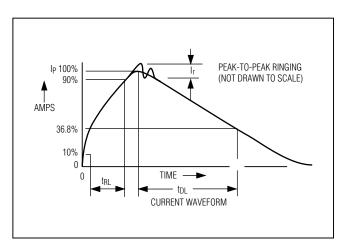


Figure 11. Human Body Current Waveform

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IEC 61000-4-2

The IEC 61000-4-2 standard covers ESD testing and performance of finished equipment. However, it does not specifically refer to integrated circuits. The MAX14775E/MAX14776E help in designing equipment to meet IEC 61000-4-2 without the need for additional ESD protection components.

The major difference between tests done using the HBM and IEC 61000-4-2 is higher peak current in IEC 61000-4-2 because series resistance is lower in the IEC 61000-4-2 model. Hence, the ESD withstand voltage measured to IEC 61000-4-2 is generally lower than that measured using the HBM. Figure 12 shows the IEC 61000-4-2 model and Figure 13 shows the current waveform for IEC 61000-4-2 ESD Contact Discharge Test.

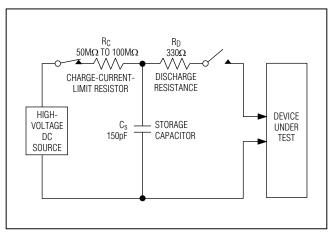


Figure 12. IEC 61000-4-2 ESD Test Model

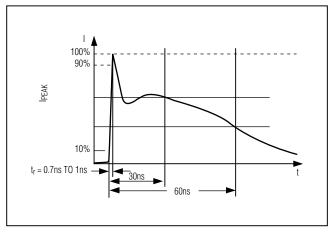
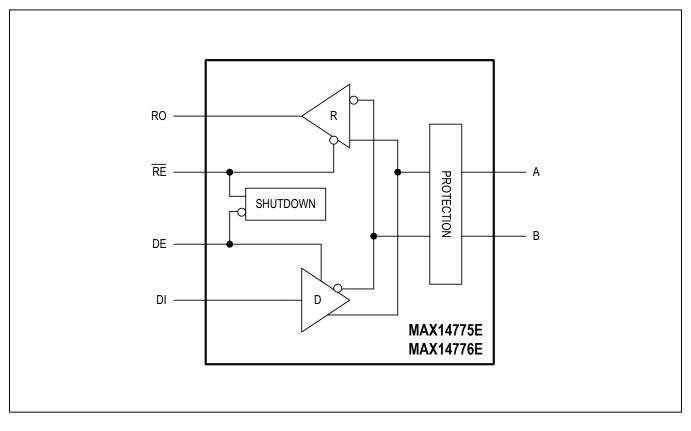


Figure 13. IEC 61000-4-2 ESD Generator Current Waveform

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Functional Diagram



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Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX14775EASA+	-40°C to +125°C	8 SOIC
MAX14775EASA+T	-40°C to +125°C	8 SOIC
MAX14775EATA+	-40°C to +125°C	8 TDFN-EP
MAX14775EATA+T	-40°C to +125°C	8 TDFN-EP
MAX14776EASA+	-40°C to +125°C	8 SOIC
MAX14776EASA+T	-40°C to +125°C	8 SOIC
MAX14776EATA+	-40°C to +125°C	8 TDFN-EP
MAX14776EATA+T	-40°C to +125°C	8 TDFN-EP

+Denotes a lead(Pb)-free/RoHS-compliant package. T = Tape and Reel

Chip Information

PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8 SOIC	S8+4	<u>21-0041</u>	<u>90-0096</u>
8 TDFN-EP	T833+2	<u>21-0137</u>	<u>90-0059</u>

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	9/16	Initial release	—

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